



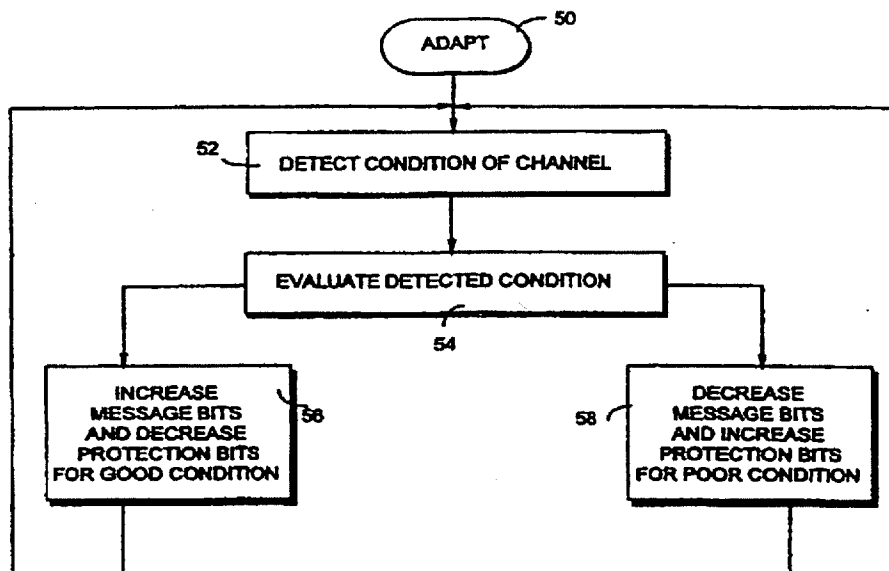
## INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

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<b>(21) International Application Number:</b> PCT/US97/03022 <b>(22) International Filing Date:</b> 26 February 1997 (26.02.97) <b>(30) Priority Data:</b> 08/610,229      4 March 1996 (04.03.96)      US <b>(71) Applicant:</b> ERICSSON INC. [US/US]; 7001 Development Drive, P.O. Box 13969, Research Triangle Park, NC 27709 (US). <b>(72) Inventors:</b> MAKOVICKA, Thomas, John; 704 Loch Highlands Drive, Raleigh, NC 27606 (US). RYDBECK, Nils, Rutger, 202 Rutherglen, Cary, NC 27511 (US). <b>(74) Agent:</b> LASTOVA, John, R.; Nixon & Vanderhye P.C., 8th floor, 1100 North Glebe Road, Arlington, VA 22201-4714 (US).		<b>(81) Designated States:</b> AL, AM, AT, AU, AZ, BA, BB, BG, BR, BY, CA, CH, CN, CU, CZ, DE, DK, EE, ES, FI, GB, GE, HU, IL, IS, JP, KE, KG, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, MD, MG, MK, MN, MW, MX, NO, NZ, PL, PT, RO, RU, SD, SE, SG, SI, SK, TJ, TM, TR, TT, UA, UG, UZ, VN, ARIPO patent (GH, KE, LS, MW, SD, SZ, UG), Eurasian patent (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM), European patent (AT, BE, CH, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE), OAPI patent (BF, BJ, CF, CG, CI, CM, GA, GN, ML, MR, NE, SN, TD, TG).  <b>Published</b> <i>With international search report.          Before the expiration of the time limit for amending the claims and to be republished in the event of the receipt of amendments.</i>

**(54) Title:** DIGITAL COMMUNICATION SYSTEM FOR ADAPTING COMMUNICATIONS PROTOCOL BASED ON A CURRENT COMMUNICATION CHANNEL CONDITION

**(57) Abstract**

A digital communications format is optimally adapted to a current communications channel condition. A message throughput rate of substantive message bits transmitted per unit time varies to add or remove message protection bits as the current channel condition changes. However, the total number of message bits and message protection bits transmitted per unit of time does not change. A current quality of the communications channel between two communications devices is determined. A ratio of substantive message bits to message protection bits is varied to adapt to that determined current channel quality. The ratio is larger for higher quality communications channels than for lower quality communications channels.



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**DIGITAL COMMUNICATION SYSTEM FOR  
ADAPTING COMMUNICATIONS PROTOCOL BASED ON  
A CURRENT COMMUNICATION CHANNEL CONDITION**

5                    **FIELD OF THE INVENTION**

The present invention relates to a system and method for adapting communications protocol to current communications channel conditions, and more particularly, to a system and method for adapting a ratio of message bits to message protection bits based on a current communications  
10 channel condition.

**BACKGROUND AND SUMMARY OF THE INVENTION**

Transmission of voice by digital techniques is now widespread, particularly in long distance telephone and digital radio telephone  
15 applications. There are two general approaches to digital voice communications. Voice waveform digitization methods take samples of the speech waveform and represent the sampled waveform amplitudes by digital, binary-coded values. At the receiver, the digital signals are converted back to analog form in an attempt to reconstruct the original  
20 speech waveform. Vocoder methods make no attempt to preserve the original speech waveform. Instead, the input speech is analyzed in terms of standardized speech features, each of which can be transmitted in digital coded form. At the receiving end, these features are reassembled, and an

output speech signal is synthesized. Ideally, the synthesized signal, as it is perceived by the human ear, closely resembles the original speech.

No matter what approach is used, errors enter the digital bitstream  
5 during transmission over a communications channel caused by noise and transmission system impairments. The number of bits which are erroneously transmitted per unit time is termed the bit error rate (BER). In radio frequency communications, and in particular mobile radio communications, higher bit error rate conditions may be caused by low  
10 signal to noise ratio, multi-path fading, phase hits, adjacent channel interference, and other (often changing) impairment characteristics of the current RF channel.

Because errors compromise the message data and in some cases  
15 render it useless, procedures have been developed to detect and correct transmission errors. These procedures invariably involve adding overhead "protection bits" to the substantive "message bits" where a message includes speech and digital data. There are a variety of error detection codes, (sometimes characterized as parity check codes) such as cyclic  
20 redundancy check codes (CRC). Other codes termed "forward error correcting codes" permit detection and correction of errors by the receiver without retransmission. Convolutional codes and block codes are examples of forward error correcting codes.

In conventional digital communications systems, the number of protection bits is fixed in an initial protocol configuration based on certain design assumptions about conditions of the communications channels over which the digital information will be transmitted. Invariably, the assumptions underlying the fixed design assumptions are not optimal for the actual condition of an individual communications channel. For example, for a poor quality communications channel, a fixed number of protection bits is likely to be too few and therefore results in a low quality received signal. Alternatively, under good channel conditions, the message information may be overprotected or over-coded with the fixed number of protection bits resulting in unnecessary lower data throughput.

What is needed is a mechanism which adjusts the ratio of bits allocated for a message signal of interest to bits allocated for protecting that signal of interest depending upon the current condition of the communications channel. Ideally, for a low quality channel, (e.g., one which results in a higher bit error rate), the number of bits allocated for the message signal of interest would be reduced, and the number of protection bits, such as those used for error detection and/or correction, would be increased. Alternatively, for higher quality channels, (e.g., one which results in a lower bit error rate condition), the number of bits allocated for the message signal would be increased, and the number of protection bits would be decreased. In addition, it is desirable to achieve this optimized

ratio of bits without affecting the data rate of transmission over the communications channel. i.e., the number of bits transmitted per second should remain constant.

- 5           The present invention solves the problems and achieves the desired objects described above by optimally adapting the digital communications format to the current communications channel condition. Although a message throughput rate of substantive message bits transmitted per unit time varies to add or remove message protection bits as the current
- 10 condition of the communications channel varies, a constant bit rate of transmission over the communications channel is maintained. In other words, the number of the substantive message bits transmitted per unit of time varies relative to a number of message protection bits transmitted per unit of time depending on the current condition of the communications
- 15 channel; but the total number of message and message protection bits transmitted per second does not change.

In a first embodiment, a digital bit stream is transmitted over a digital communications channel between first and second communications

20 devices. The bit stream is divided into a first variable number of bits allocated to substantive message information and a second variable number of bits allocated to protecting the substantive message information. The total of the first and second variable number of bits transmitted per unit of

time is substantially constant. A current quality of the communications channel between the two communications devices is determined preferably throughout the communications. The current channel quality can be detected using a variety of methods including for example current bit error rate, received signal strength, etc. The ratio of the first and second variable numbers of bits may can be continuously varied to any value to adapt to the current quality communications channel. Alternatively, a plurality of ratios of the first variable number and the second variable number may be established. If a first ratio is larger than a second ratio, the first ratio is selected for the higher quality communications channel, and the second ratio is selected for a lower quality communications channel.

The present invention finds particular application in a second embodiment to radio frequency (RF) digital communication systems such as cellular radio telephone and public safety trunk (PST) radio systems. Each portable radio includes an RF transmitter with a voice coder (vocoder) and an RF receiver for conducting digital voice communications over an RF communications channel at a fixed bit rate. The RF communications channel may be divided into a series of frames with each frame including the first number of bits allocated to the speech carried by that frame and a second number of coding bits allocated to encode those speech bits in the frame. One or more base stations are provided to establish and coordinate RF communication between two or more of the portable radios.

A channel condition detector (which may be located in each of the radios and/or a cell/site base station) detects the current condition of the RF communications channel. A radio controller varies the first and second number of bits based on the detected condition of the communications channel while maintaining a substantially constant bit rate. More bits are allocated to the first number of speech bits in the frame, and fewer bits are allocated to the second number of coding bits in the frame when the channel condition is more optimal. When the channel condition is less optimal, fewer bits are allocated to the first number of speech bits in the frame, and more bits are allocated to the second number of coding bits in the frame.

In an alternative embodiment, the radio controller varies the first and second number of bits based on the base station site/cell in which the radio is currently located. Certain base station sites/cells preset the numbers of first and second bits. As the radio roams between sites/cells, the radio changes the first and second number of bits to match the cell's/site's preset numbers.

The features, objects, and advantages of the present invention will become more apparent from the detailed description set forth below taken in conjunction with the drawings in which like reference characters identify corresponding elements throughout.



### BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 is a schematic block diagram illustrating the basic components of a mobile/portable radio communications system;

5

FIG. 2 is a schematic view illustrating a cell or site pattern for the radio communications system shown in FIG. 2;

FIG. 3 is a flow chart diagram outlining adaptive procedures followed by radios in accordance with the present invention;

10

FIG. 4 is a function block diagram of a radio unit which may be employed to implement the present invention;

15

FIG. 5 is a diagram showing a digital communications channel having a framed format;

FIGS. 6 and 7 are high level function block diagrams illustrating various embodiments of the present invention;

20

FIG. 8 is a function block diagram showing an example for implementing the present invention; and

FIGS. 9A and 9B illustrate another example embodiment in accordance with the present invention.

### **DETAILED DESCRIPTION OF PREFERRED EMBODIMENTS**

5 In the following description, for purposes of explanation and not limitation, specific details are set forth, such as particular diagrams and techniques in order to provide a thorough understanding of the present invention. However, it will be apparent to one skilled in the art that the present invention may be practiced in other embodiments and in other ways  
10 that depart from these specific, illustrative details. In other instances, detailed descriptions of well known methods, devices, and circuits are omitted so as not to obscure the description of the present invention with unnecessary detail.

15 FIG. 1 illustrates a mobile/portable radio communications system such as is employed in cellular radio telephone and public safety trunked (PST) radio communications. Plural portable/mobile radio units 21 roam throughout a communications area divided into a plurality of cells or sites represented for example as hexagons 36. Each cell/site includes a base  
20 station 23 which includes a plurality of radio frequency transceivers for communicating over radio frequency communications channels with the mobile/portable radio units 21. Each base station 23 is connected to a central switch and control 31 which provides an interface to the public

switched telephone network (PSTN) 30 and optionally to a dispatch console 34 for PST-type applications. Of the number of base station radio frequency channels, one is designated as a control channel for transmitting control information such as identification, paging, call request, call set-up, status, as well as other information. The remaining RF channels may be referred to as working channels and are allocated to handle both voice and data calls.

Because the radio units 21 are portable, their position relative to a current "home" base station 23 changes. That relative position may change considerably in a short period of time when the radio unit is transported in an automobile. FIG. 2 illustrates two mobile type radios 21' and 21" separated from their home base station 23 by different distances. Radio 21' is relatively close to the base station and therefore typically receives a stronger, higher quality signal than mobile radio 21" which is further away from base station 23. Of course, other factors in addition to distance may effect the quality of communication including buildings and other obstructions, weather conditions, and interference from other radios. Nevertheless, assuming for purposes of illustration that radio 21' has a better channel than radio 21", fewer bits are needed to protect digital messages transmitted from radio 21'.

A first general embodiment for adapting to current channel conditions is illustrated in the adapt routine (block 50) illustrated in the flowchart of FIG. 3. The current condition of the communications channel is detected (block 52) at the beginning and preferably (but not necessarily) throughout the duration of the communication. The detected condition is evaluated (block 54) to determine if there has been any change in the previously detected condition of the channel. By frequently or substantially continuously detecting the condition of the channel, the system can quickly adapt to situations where a good channel quickly becomes a bad channel, e.g., where the radio moves underneath a bridge or behind a building. If the channel is in good condition, the number of message bits may be increased, and the number of protection bits may be decreased (block 56). Alternatively, if the channel condition is poor or otherwise has deteriorated, the number of message bits may be decreased, and the number of protection bits may be increased (block 58).

The present invention may be applied to any type of digital communications system using any communications media including telephone wires, coaxial cables, optical fibers, and radio frequency channels. Similarly, the present invention may be employed by any type of transceiver suitable for digital communication over any of these communications media. However, a particularly advantageous application of the present invention is to digital radio communications. For this

application, FIG. 4 illustrates in function block format a suitable radio transceiver which may be employed for implementing the present invention in a portable/mobile radio. While this illustrative embodiment is described in terms of being performed by a mobile/portable radio, the present  
5 invention may also be implemented at a base station transceiver for adapting its communications with other units, some of which may be moving. Similarly, the present invention could be implemented in wireline digital communications devices that employ modems and vocoders.

10 Radio frequency signals are transmitted and received by way of antenna 140 through conventional combiner/diplexer 142 which permits reception and transmission using the same antenna. A radio receiver 144 receives signals including both speech and data. Speech signals are routed to voice coder/decoder (vocoder) and other audio signal processing  
15 circuits 152 which decodes the received digital information into an analog voice signal for generation at an audio speaker. Digital data is routed to conventional modem 154 for demodulation and connection to an appropriate digital interface. Outgoing data is appropriately encoded and modulated in modem 154 and forwarded to RF transmitter 146 for  
20 transmission via combiner 142 and antenna 140.

Speech is detected by a suitable microphone (MIC) which may be voice-activated or may be enabled in PST-type systems by a push to talk

(PTT) button for transmission-trunked communications. The received voice is processed and vocoded by block 152 and passed to RF transmitter 146 for transmission via combiner 142 and antenna 140. Microprocessor 156 with memory 158 coordinates and controls the operation of the various circuitry via an input/output (I/O) interface 150. A display and keypad for operator interface and control of the radio is also provided via I/O 150. A received signal strength indicator (RSSI) 148 is connected to the received signal output from combiner 142 for detecting received signal strength. Received signal strength is one indicator of the current channel condition and may be used alone or in combination with other factors like BER to evaluate the current communications channel condition.

Typically, digital communications are formatted using sequential frames (1, 2, 3, 4, 5, ...) as shown in FIG. 5. Each frame includes a predetermined total number of bits. In this example, message bits are illustrated as being in one portion of each frame while message protection bits, such as cyclic redundancy code (CRC) bits, are allocated to a second portion of the frame. Alternatively, the message bits could be encoded over the entire frame with there being no division between message bits and message protection bits, e.g., as with various voice coding approaches including convolutional and block codes. However, even using such message coding approaches, a certain number of bits in the frame are allocated to the message bits actually conveyed in that frame with the

remaining bits being as protection bits. The present invention applies to any type bit allocation and/or frame formatting technique.

As described above, the radio detects the current condition of the  
5 channel such as the received signal strength which is routed via I/O 150 to radio microprocessor 156 for analysis. Of course, other channel condition parameters could be used. For example, channel quality could be determined using residual or average bit error rate determinations measured by CRC techniques, Viterbi demodulation metrics, speech quality  
10 measurements, etc. The microprocessor 156 then determines based on this detected condition of the channel whether the current allocation of bits between the message and protection bits is optimal. The microprocessor 156 then selects the appropriate number of bits used to encode the data or voice code the speech to be transmitted by the radio.

15

The present invention and alternative embodiments need not be dependent on detected channel condition. For example, an operator may manually select appropriate bit allocations independent of the detected channel condition. Alternatively, each base station may predetermine the  
20 bit allocations to be used by radios currently located in its geographical site/cell. Roaming radios would switch to the predetermined bit allocations for each newly entered site/cell area when roaming into each area. The predetermined bit allocations could be broadcast by the base station, or they

could be stored in a look up table in each radio along with a site/cell identification. Then when the radio roams into the new site/cell and detects the broadcast site/cell identification, the radio can look up the appropriate predetermined bit allocations for that site/cell.

5

One way of managing these numbers of bits is to establish a ratio of bits allocated for the message itself to bits allocated for message protection. A plurality of such ratios A, B, ... N may be established with the microprocessor 156 selecting the ratio appropriate for the current channel  
10 condition. For example, under more optimal channel conditions, the ratio is increased or a larger ratio is selected. In less optimal conditions, the ratio is decreased or a smaller ratio is selected.

Even though the allocation of bits to message and message  
15 protection varies which affects message throughput, (i.e., the number of message bits transmitted per second), the actual bit rate in terms of total bits (message + protection) transmitted per second is substantially fixed. In this way, system performance is optimized in accordance with the current condition of the channel. For example, for a current channel with high bit  
20 error rate condition, a lower ratio is used. In a vocoding environment, this translates into a lower rate speech coder because fewer bits of speech information are actually transmitted per unit time than would be otherwise. The present invention may be used with any speech encoding algorithm



including for example linear predictive coding. Using a greater number of bits are used to vocode the speech increases the chance that the speech can be decoded intelligibly by the receiver despite the poor channel.

Accordingly, speech quality improves with additional bits allocated or used  
5 in speech coding. Alternatively, speech with reasonable quality can be intelligibly decoded with fewer vocoding bits under better channel conditions thereby increasing the throughput rate of the speech data transmitted per unit time. A higher rate speech coder is effectively implemented when channel conditions are improved.

10

Accordingly, the performance and reliability of the communications system are optimized for the currently existing communications conditions. In essence, the throughput rate of the message (either speech bits or data bits) varies to add or remove overhead bits such as forward error  
15 detection/correction bits as the communications channel changes. Another advantage of the present invention, in addition to improved performance and reliability, is that substantially constant bit error rate may be achieved independent of the channel condition.

20 Bit allocation or ratio selection can be performed at either the transmitting radio or the receiving radio at the initiation of the communication and preferably throughout the communication. Bit allocation/ratio selection just at the start of the communication may be

suitable for relatively stationary communications or in environments where a channel is only used for a short period of time such as in half-duplex, transmission-trunked systems.

5        Detection/determination of the signal channel condition  
detection/determination and the allocation of bits can be implemented in  
any number of ways. FIGS. 6 and 7 show two, non-limiting, example  
implementations. In FIG. 6, receiving transceiver 1 determines the ratio of  
message to message protection bits based on detected signal quality of the  
10 received signal over the information forward link. The detected signal  
quality is then communicated back to the transmitting transceiver via a  
return link. In the cellular and PST communications environments which  
employ control and working channels as described earlier, the signal  
quality could be detected and/or communicated using either the control  
15 channel or a working channel. Based on the detected signal quality  
returned to the transmitting transceiver 2, a ratio of message-to-protection  
bits may be selected, e.g., ratio A, ratio B, ... ratio N. In this  
implementation, the ratio is selected effectively by the receiving  
transceiver 1.

20

Alternatively as shown in FIG. 7, the ratio could be selected by the  
transmitting transceiver 2. A signal quality detector in transmitting  
transceiver 2 detects the quality of the current communications condition

between transceiver 1 and transceiver 2 over a return link. Based on the detected signal quality, one of the possible ratios A, B, ... N is selected. This configuration, of course, assumes some correlation in channel conditions from the transmitter to receiver and the channel conditions from  
5 the receiver to the transmitter. The receiving transceiver 1 determines at what ratio the information has been transmitted over the forward link. This information can be transmitted as control information over the channel, or the receiving transceiver 1 may determine the ratio directly from the received information. One possible technique for determining the ratio  
10 from the received information would be to decode the information using all the ratios in which the information could have been transmitted, and then determine which ratio has the lower bit error rate or a matching CRC.

An example of variable throughput rate coding of speech is now  
15 described in conjunction with FIG. 8. This example uses a straightforward approach of pulse code modulated (PCM) speech using a simple coding scheme. An analog to digital converter (A/D) receives the analog speech and generates PCM data. At each sample of the speech, an 8-bit digital code is generated corresponding to the amplitude level of the speech  
20 waveform at that sample moment. Each 8-bit PCM input is provided to a voice coder that encodes the speech at three variable rates: 12/8, 12/6, and 12/4.

The voice coder may only encode a certain number of the most significant bits of the 8-bit PCM input. In this example, the three most significant bits are encoded into seven bits with the remaining least five significant bits being uncoded. Using this 7/3 most significant bit code, the  
5 eight input bits are converted to twelve coded output bits. In other words, there are eight message bits and four protection bits for twelve total bits. The lower rate coders 12/6 and 12/4 code the most significant bits more heavily (with more protection bits) to protect those bits under less optimal communications conditions. For example, the 12/4 code may be achieved  
10 by simply repeating the four most significant bits three times for a total of twelve bits. Regardless of the particular coding ratio selected, a constant twelve bits are transmitted by the transmitter over the channel. Thus, the data bit rate per second is fixed while the data throughput rate is variable. In this example, anywhere from eight to four speech bits may actually be  
15 transmitted during that time interval.

At the receiver, the voice decoder decodes the received twelve bits using the appropriate decoding algorithm in order to extract the corresponding number of speech bits, i.e., 8, 6, or 4 from the twelve total  
20 bits. For the 12/4 encoded bits, for example, the 12/4 decoder might decode the received twelve bits by majority voting bits (0-3), (4-7), and (8-11) to arrive at the four actual speech bits. The digital analog converter then formulates an analog waveform that approximates this portion of the

original speech waveform. When the received speech bits are less than the number of inputs into the digital to analog converter, as in this example of four speech bits being input an 8-bit digital to analog converter, the least significant bits may be filled with binary zeros.

5

FIGS. 9A and 9B show a more sophisticated vocoding scheme. In this example, the vocoder bit transmission rate is 7.2 kbps with 10 msec frames of 72 bits. FIG. 9A illustrates a bit allocation appropriate for a relatively good communications channel, i.e., one which requires low bit error protection. A total number of 45 speech bits ( $12 + 12 + 11 + 10$ ) are selectively coded using an additional 27 protection bits to generate a total of 72 output bits ( $24 + 23 + 15 + 10$ ). In terms of data rate, of the total 7.2 kbps data transmission rate, 4.5 kbps are allocated to the speech bits, and 2.7 kbps are allocated to protection bits. In this example, certain ones of the 45 speech bits are selectively encoded based on their priority or other significance. The twelve top "priority 1" bits are encoded using a (24,12) extended Golay code. The next twelve "priority 2" bits are encoded with a (23,12) extended Golay code. The next eleven "priority 3" bits are encoded using a (15,11) Hamming code, and the least significant "priority 4" bits are not encoded at all.

For a less optimal channel condition, the bit allocation ratio illustrated in FIG. 9B may be adopted. In this instance, only 37

(7 + 12 + 12 + 6) speech bits are transmitted for each total of 72 output bits transmitted over the channel per unit in time, e.g., frame. In terms of data rate, 3.7 kbps are allocated to vocode speech bits, and 3.5 kbps are allocated to forward error correction bits. Thus, the 4.5 kbps vocoder shown in FIG. 9A has a higher rate than the 3.7 kbps vocoder shown in FIG. 9B. However, the increased number of protection bits in FIG. 9B improves the chance that intelligible speech will be decoded at the receiver under the poor channel condition. For both protocol/format ratios, the same bit error rate for two different communications may be achieved even though the transmission conditions under which these communications occur may vary significantly. The seven priority 1 bits are encoded using a (19,7) extended Golay code. The next twelve priority 2 bits are encoded using a (24,12) extended Golay code; and the next twelve priority 3 bits are encoded using a (23,12) Hamming code, and the six lowest priority 4 bits are not coded.

This description of preferred embodiments is provided to enable a person skilled in the art to make and/or use the present invention. Various modifications to these embodiments will be readily apparent to those skilled in the art, and the general principles defined herein may be applied to other embodiments. Thus, the present invention is not intended to be limited to these illustrative embodiments but is to be accorded the widest

scope consistent with the principles and novel features disclosed in  
accordance with the following claims.

**WHAT IS CLAIMED IS:**

3           1.     In a digital communications system where a digital bit stream  
4     transmitted over a digital communications channel between a first  
5     communications device and a second communications device is divided  
6     into a first variable number of bits allocated to substantive message  
7     information and a second variable number of bits allocated to protecting the  
8     message information, a total number of the first and second variable  
9     number of bits transmitted per unit time interval being substantially  
10    constant, a method comprising the steps of:

11           determining a current quality of the digital communications channel,  
12    and

13           varying a ratio of the first variable number of bits to the second  
14    variable number of bits to adapt to the current quality of the digital  
15    communications channel.

1           2.     The method in claim 1. further comprising:  
2           establishing plural ratios of the first variable number to second  
3    variable number, wherein a first ratio is larger than a second ratio, wherein  
4    the varying step includes selecting the first ratio for a higher quality  
5    communications channel and the second ratio for a lower quality  
6    communications channel.



1           3.     The method in claim 1, wherein the ratio is continuously varied  
2     to adapt to the current quality of the communications channel.

3

4           4.     The method in claim 2, wherein the current channel quality is a  
5     bit error rate of the current channel and the message information is digitized  
6     speech, and wherein the first ratio is selected when the bit error rate is  
7     relatively low so that more bits may be allocated to digitized speech and the  
8     second ratio is selected when the bit error rate is relatively high so that  
9     more bits may be allocated to error detection or error correction.

1           5.     The method in claim 2, wherein the current channel quality is a  
2     received signal strength of the current channel and the message information  
3     is digitized speech, and wherein the first ratio is selected when the received  
4     signal strength is relatively strong so that more bits may be allocated to  
5     digitized speech and the second ratio is selected when the received signal  
6     strength is relatively weak so that more bits may be allocated to error  
7     detection or error correction.

1           6.     The method in claim 1, wherein the ratio is manually set by a  
2     user of one of the first and second devices irrespective of the  
3     communications channel condition.

1           7.       The method in claim 1, wherein the first and second devices  
2       are portable radio transceivers and each radio transceiver includes a voice  
3       coder for vocoding speech and decoder for decoding vocoded speech, and  
4       wherein the first variable number of bits include speech bits such that  
5       changing the ratio changes the number of bits used to vocode a frame of  
6       speech.

1           8.       The method in claim 1, wherein the second number of bits are  
2       coding bits used to encode message bits with more significant bits of the  
3       message bits being coded more than less significant message bits.

1           9.       The method in claim 1, wherein the second number of bits  
2       include block coded bits.

1           10.      The method in claim 1, wherein the second bits are  
2       convolutionally coded bits.

1           11.      The method in claim 1, wherein the second bits are repeated  
2       ones of the first bits.

1           12.      A radio frequency (RF) digital communications system that  
2       adapts to current communications conditions, comprising:

3 plural portable radios, each portable radio including an RF transmitter  
4 with a voice coder (vocoder) and an RF receiver for conducting digital  
5 voice communications over an RF communications channel at a fixed bit  
6 transmission rate, where the RF communications channel is divided into a  
7 series of frames with each frame including a first number of bits allocated  
8 to speech carried by the frame and a second number of coding bits allocated  
9 to encode the speech bits:

10 a base station including a plurality of RF transceivers for establishing  
11 and coordinating RF communications between the portable radios;

12 a channel condition detector for detecting a current condition of the  
13 RF communications channel; and

14 a controller for varying the first and second number of bits based on  
15 the detected condition of the communications channel while maintaining a  
16 substantially constant bit rate such that more bits are allocated to the first  
17 number of speech bits in the frame and fewer bits are allocated to the  
18 second number of coding bits in the frame for a first channel condition, and  
19 fewer bits are allocated to the first number of speech bits in the frame and  
20 more bits are allocated to the second number of coding bits in the frame for  
21 a second channel condition that is less desirable than the first channel  
22 condition.

1 13. The system in claim 12, wherein the controller establishes a  
2 variable ratio of the first number to second number, with a first ratio being

3 larger than a second ratio, and selects the first ratio for the first channel  
4 condition and the second ratio for the second channel condition.

1 14. The system in claim 13, wherein the channel condition  
2 corresponds to a quality of the channel and the variable ratio is  
3 continuously varied to adapt to the current quality of the communications  
4 channel.

1 15. The system in claim 13, wherein the current channel condition  
2 is a bit error rate of the current channel, and wherein the first ratio is  
3 selected when the bit error rate is relatively low and the second ratio is  
4 selected when the bit error rate is relatively high.

1 16. The system in claim 13, wherein the detected current channel  
2 condition is a received signal strength of the current channel, and wherein  
3 the first ratio is selected when the received signal strength is relatively  
4 strong and the second ratio is selected when the received signal strength is  
5 relatively weak.

1 17. The system in claim 13, wherein the detected current channel  
2 condition is a demodulation parameter determined using a plurality of  
3 channel condition measurement variables.

1           18.    The system in claim 13, wherein the first and second numbers  
2           of bits are manually set by a user of one of the portable radios irrespective  
3           of the communications channel condition.

1           19.    The system in claim 13, wherein the first and second numbers  
2           of bits are set by the base station independent of the communications  
3           channel condition.

1           20.    The system in claim 12, wherein the channel condition detector  
2           is activated to detect the current channel quality when the portable radio is  
3           receiving over the communications channel.

1           21.    The system in claim 12, wherein the base station and radios  
2           communicate control information over an RF control channel in addition to  
3           the RF communications channel, the detector detecting a current channel  
4           condition of the RF control channel, and wherein the controller determines  
5           whether to vary the first and second number of bits on the RF  
6           communications channel based on the current channel condition of the RF  
7           control channel.

1           22.    The system in claim 12, wherein the coding was block codes or  
2           convolutional codes.

1           23.    The system in claim 21, wherein the base station includes the  
2   channel condition detector and controller.

1           24.    A method for adapting a digital communications format to a  
2   current communications channel condition wherein a message throughput  
3   rate of substantive message bits transmitted per unit time varies to add or  
4   remove message protection bits as the current condition of the  
5   communications channel varies while maintaining a substantially constant  
6   bit rate.

1           25.    The method in claim 24, wherein a number of the substantive  
2   message data bits transmitted per unit time varies relative to a number of  
3   message protection bits transmitted per unit time depending on the current  
4   condition of the communications channel.

1           26.    The method in claim 24, wherein the protection bits are used  
2   to encode message bits with more significant ones of the message bits being  
3   more heavily protected than less significant message bits.

1           27.    The method in claim 24, wherein the protection bits are block  
2   coded bits.

1           28.    The method in claim 24, wherein the protection bits are  
2   convolutionally coded bits.

1           29.    The method in claim 24, wherein the message bits represent  
2   voice.

1           30.    The method in claim 24, wherein the current communications  
2   channel condition is determined from bit error rate determined using CRC  
3   techniques, Viterbi demodulation metrics, or speech quality measurements.

1           31.    A radio frequency (RF) digital communications system,  
2   comprising:  
3       plural portable radios, each portable radio including an RF transmitter  
4   with a voice coder (vocoder) and an RF receiver for conducting digital  
5   voice communications over an RF communications channel at a fixed bit  
6   transmission rate, where the RF communications channel is divided into a  
7   series of frames with each frame including a first number of bits allocated  
8   to speech carried by the frame and a second number of coding bits allocated  
9   to encode the speech bits;

10       a first base station corresponding to a first geographical coverage area  
11   and including a plurality of first RF transceivers for establishing and  
12   coordinating a first RF communication between portable radios, at least one  
13   of the portable radios involved in the first RF communication being located

14 in the first geographical area, where the first base station establishes a first  
15 format for the first number of bits and the second number of bits for the  
16 first RF communication such that more bits are allocated to the first number  
17 of speech bits in the frame and fewer bits are allocated to the second  
18 number of coding bits; and  
19 a second base station corresponding to a second geographical  
20 coverage area and including a plurality of second RF transceivers for  
21 establishing and coordinating a second RF communication between  
22 portable radios, at least one of the portable radios involved in the second  
23 RF communication being located in the second geographical area, where  
24 the second base station establishes a second format for the first number of  
25 bits and the second number of bits for the second RF communication such  
26 that fewer bits are allocated to the first number of speech bits in the frame  
27 and more bits are allocated to the second number of coding bits.

1 32. The system in claim 31, wherein when one of the portable  
2 radios enters the first geographical area, the one portable radio adopts the  
3 first format for communications with the first base station, and when the  
4 one portable radio enters the second geographical area, the one portable  
5 radio adopts the second format for communications with the second base  
6 station.



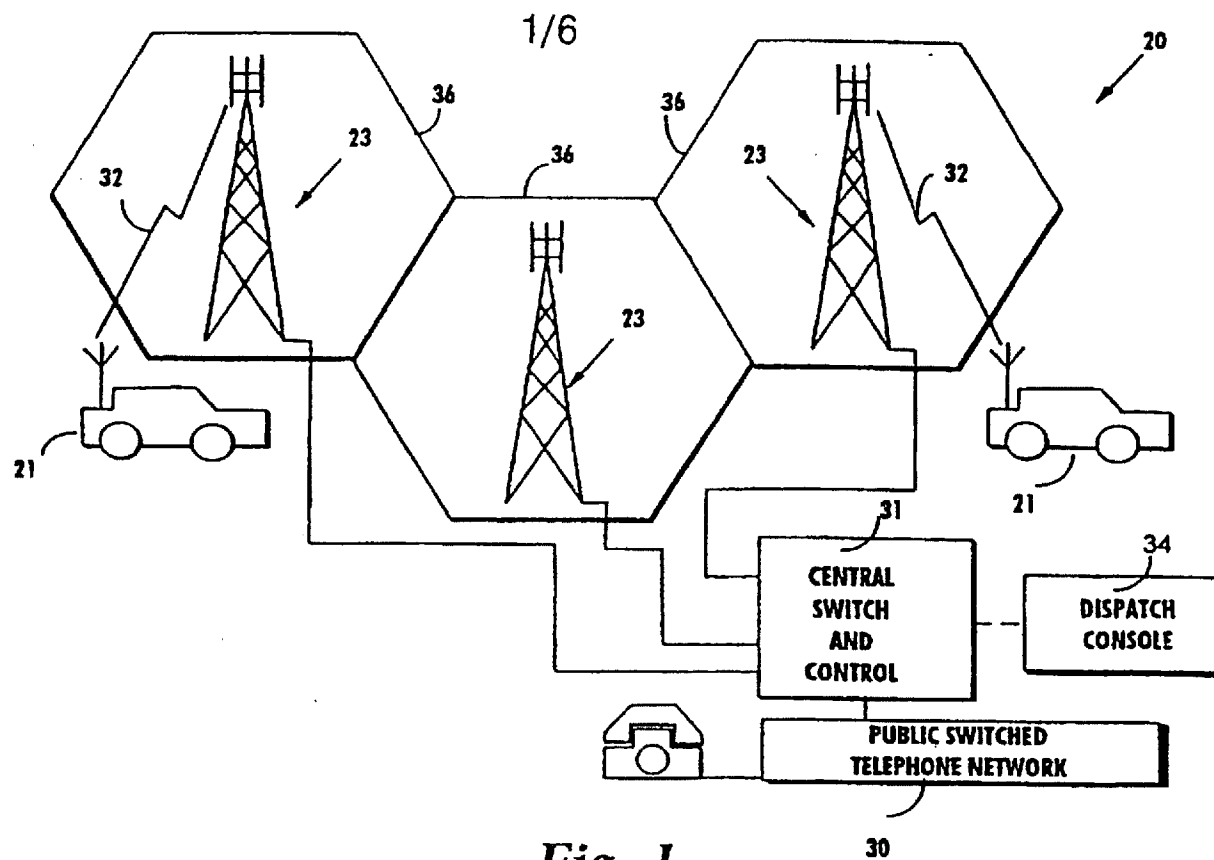


Fig. 1

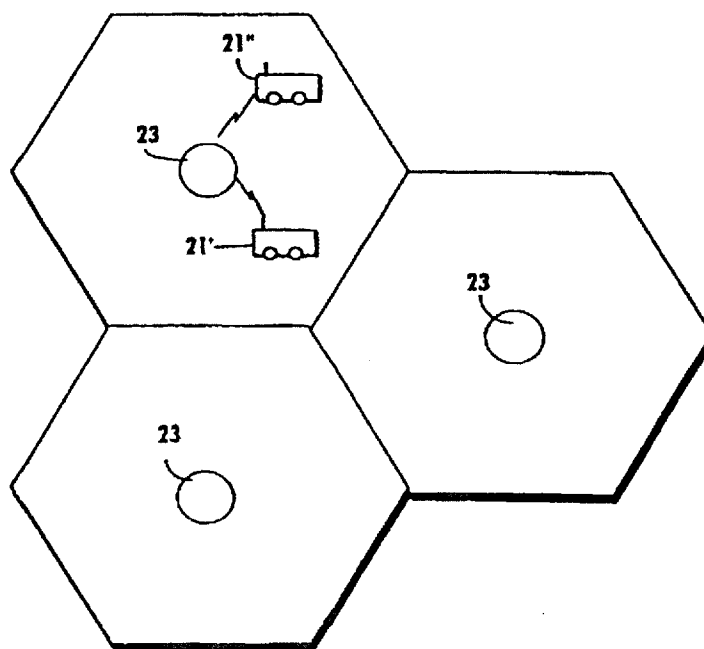


Fig. 2

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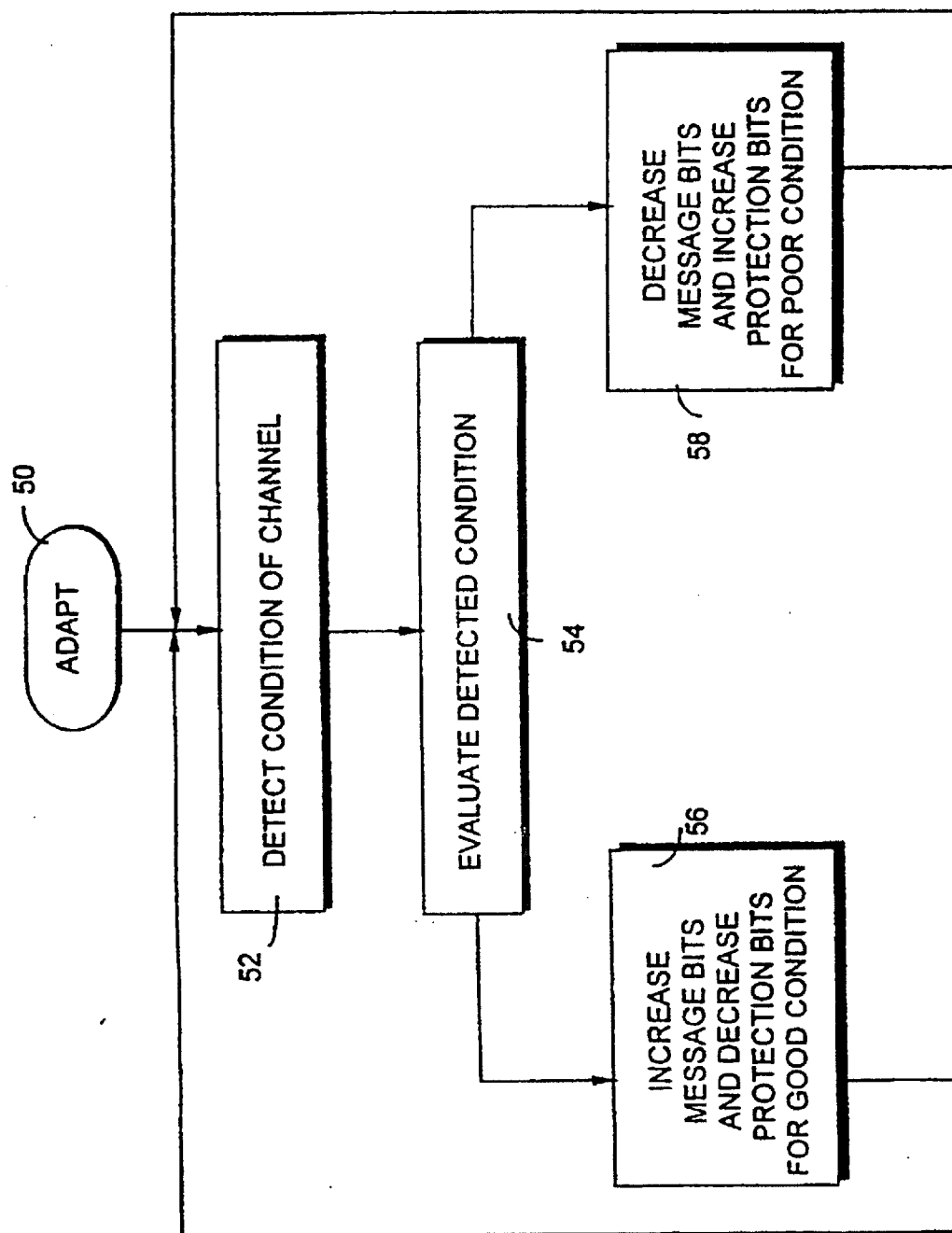


Fig. 3

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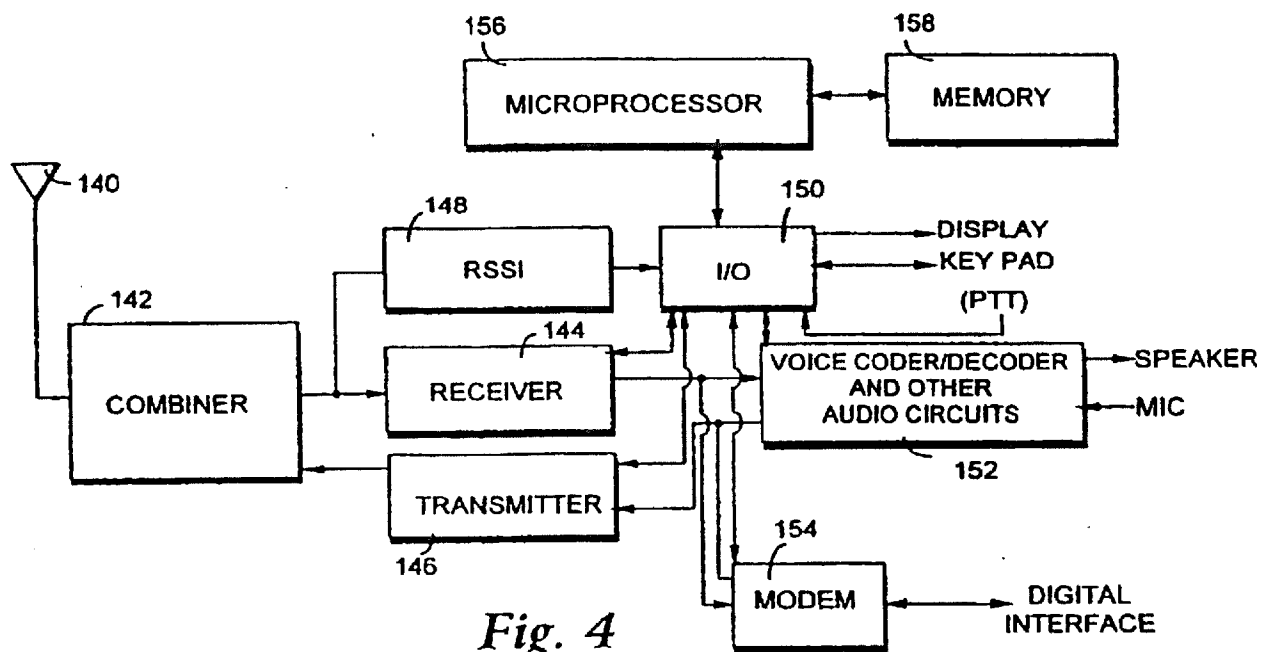


Fig. 4

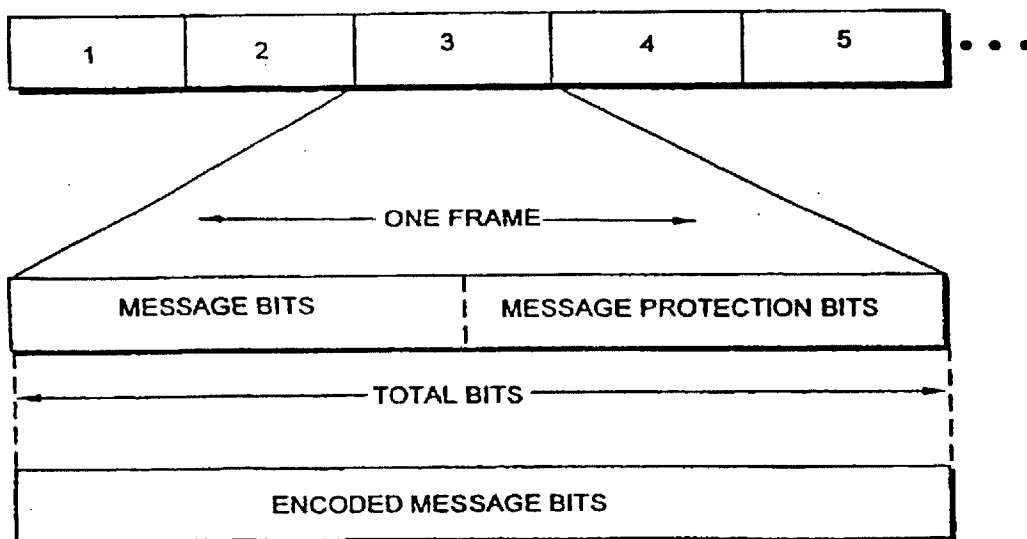
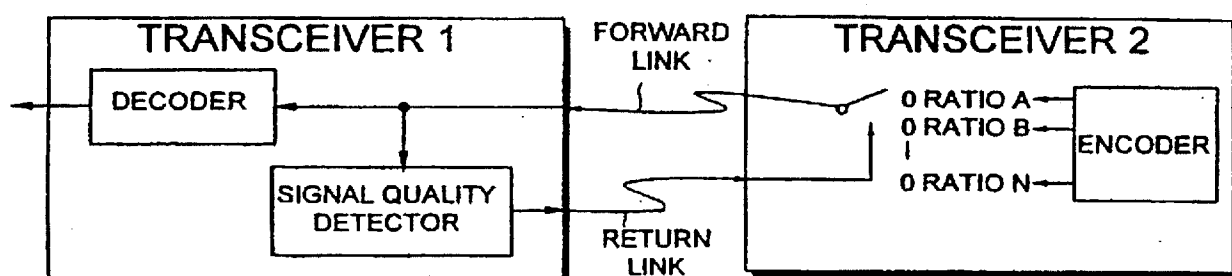
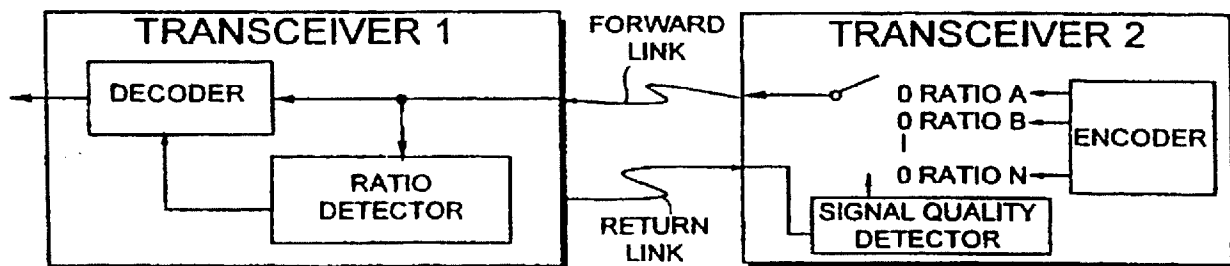


Fig. 5

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*Fig. 6**Fig. 7*

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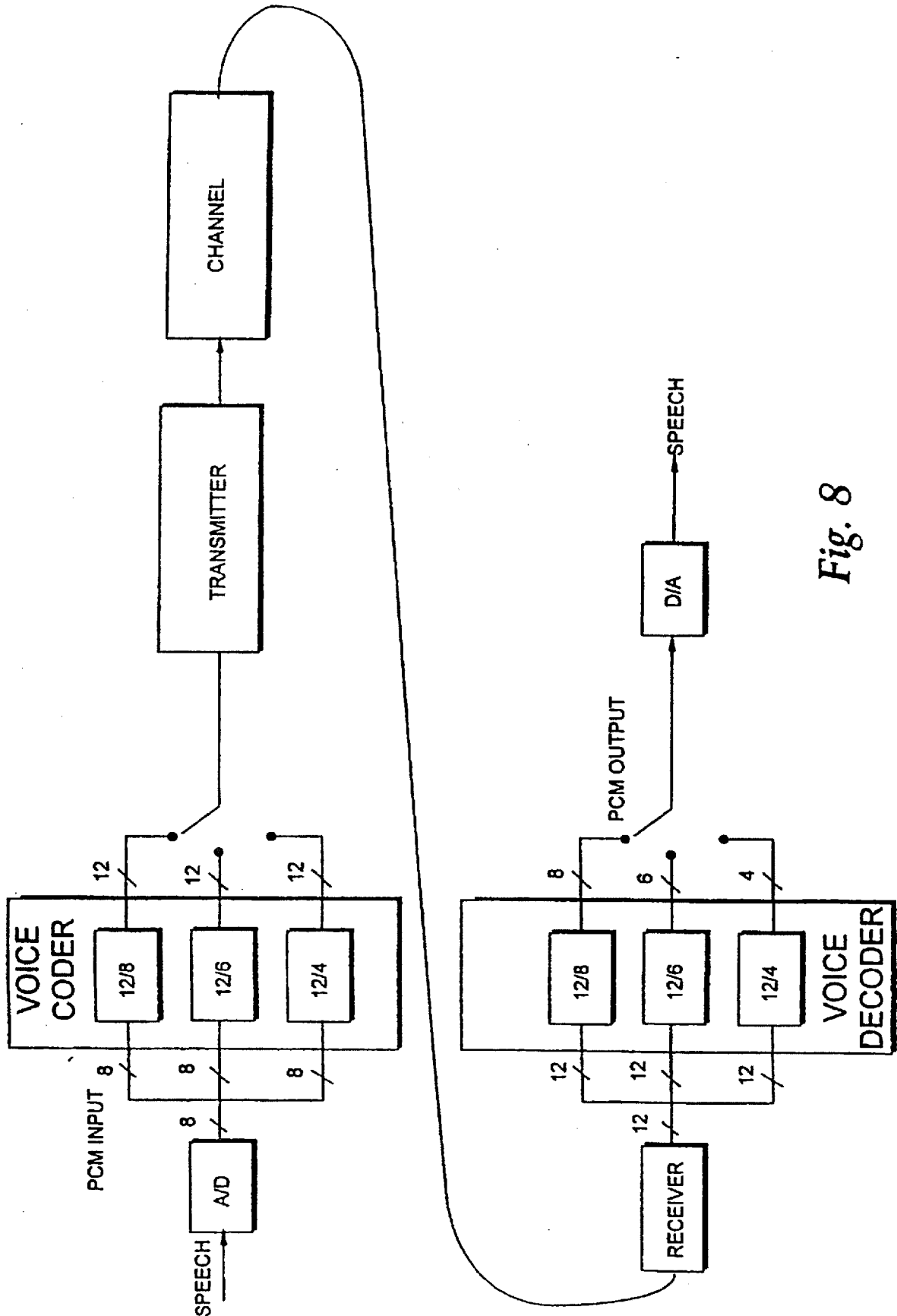
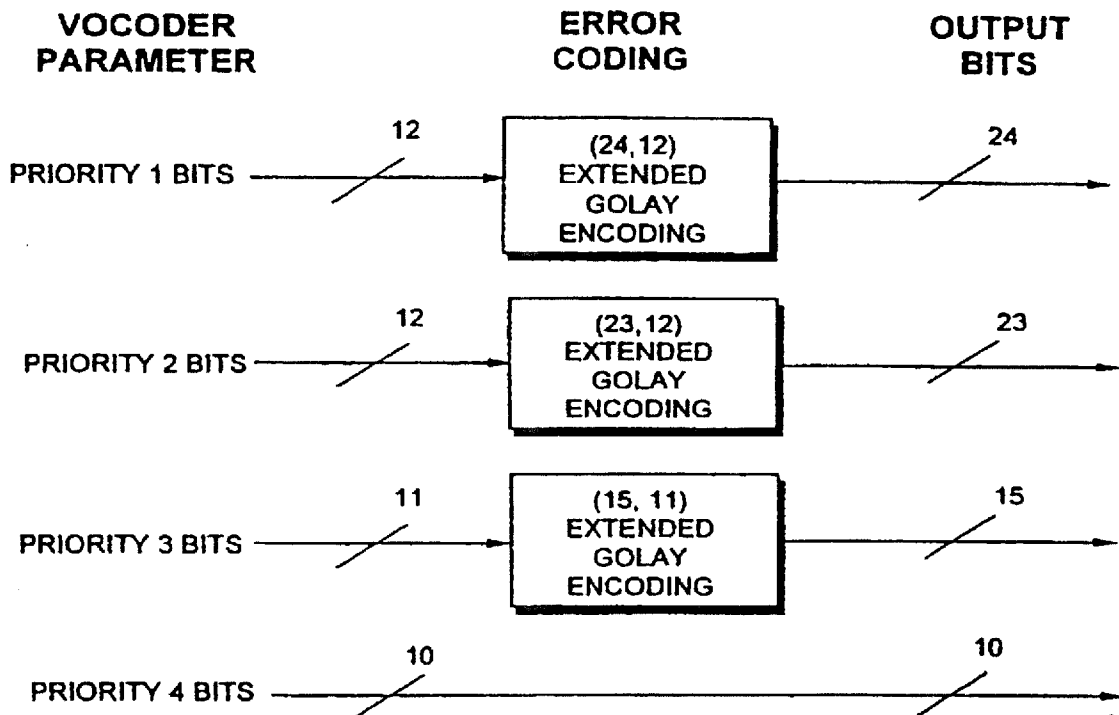
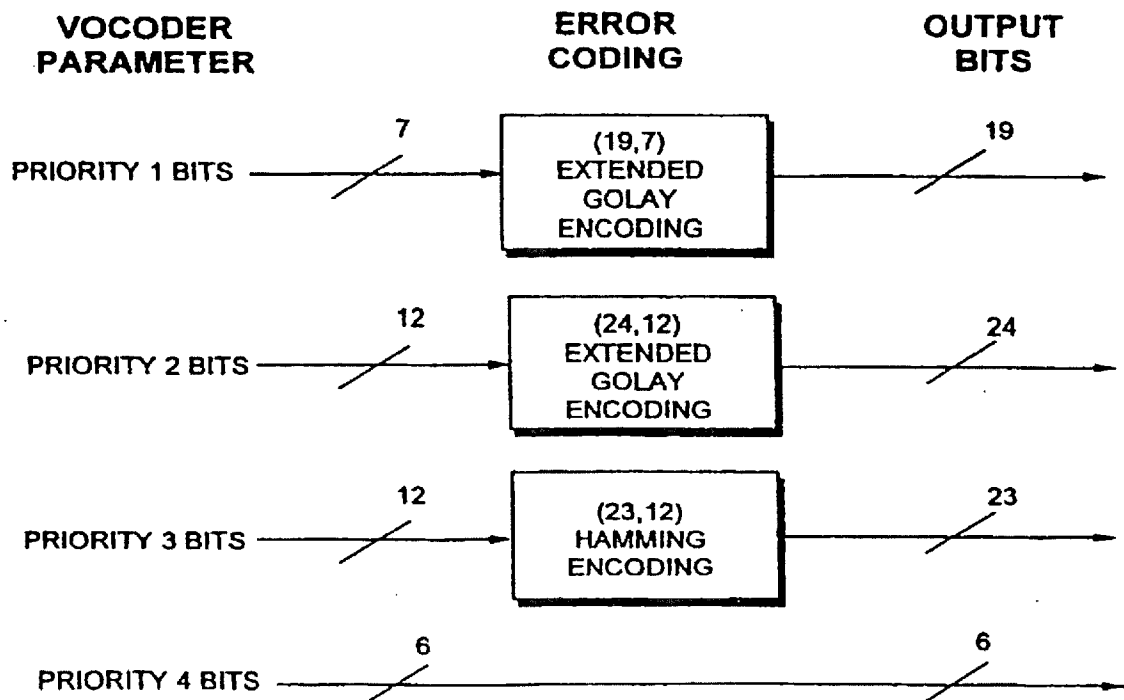


Fig. 8

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*Fig. 9a**Fig. 9b*

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# INTERNATIONAL SEARCH REPORT

Internat. Application No  
PCT/US 97/03022

A. CLASSIFICATION OF SUBJECT MATTER  
IPC 6 H04L1/20 H04L1/12 H04B7/26

According to International Patent Classification (IPC) or to both national classification and IPC

## B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 6 H04L H04Q

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

## C. DOCUMENTS CONSIDERED TO BE RELEVANT

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X	DE 34 43 041 A (SIEMENS AG) 28 May 1986	1,2, 8-10, 24-28,30
Y	see page 5, line 4 - line 13 see page 6, line 7 - line 13 --- -/--	16,22

☒ Further documents are listed in the continuation of box C.

☒ Patent family members are listed in annex.

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Date of the actual completion of the international search

2 July 1997

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